

# Lecture 10: Fourier Analysis and Frequency Response

Tuesday, November 10, 2020

8:58 AM

## References:

Tranquillo JV. *Biomedical Signals and Systems*, Morgan & Claypool Publishers, Dec. 2013. Ch. 8 & Ch. 11 (Sec. 8.5; Sec. 11.1 - 11.4).

Laplace vs Fourier revisited (see Lecture 3):

$$\mathcal{L}\{x(t)\} = \int_0^{+\infty} x(t) \cdot e^{-st} dt = X(s)$$

$$\mathcal{F}\{x(t)\} = \int_{-\infty}^{+\infty} x(t) \cdot e^{-j\omega t} dt$$

$$= \mathcal{L}\{x(t)\} \text{ where: } \begin{cases} s = j\omega \quad (\sigma=0) \\ x(t) = 0 \text{ for all } t \leq 0 \\ \text{I.C.} \equiv 0 \end{cases}$$

$$\text{If I.C.} = 0: \quad X(s) = H(s) \cdot f(s)$$

$$X(j\omega) = H(j\omega) \cdot f(j\omega)$$

Transfer function in  
Laplace and Fourier domains  
( $s = j\omega$ )

Main differences between Laplace and Fourier (see Practice Quiz 1):

$$1) \quad \begin{array}{l} \mathcal{L} : 0 \leq t \leq +\infty \quad + \text{I.C. @ } 0 \\ \mathcal{F} : -\infty \leq t \leq +\infty \quad \text{No I.C.} \end{array}$$

$$2) \quad H(s) = H(j\omega) \quad \text{where } s = j\omega$$

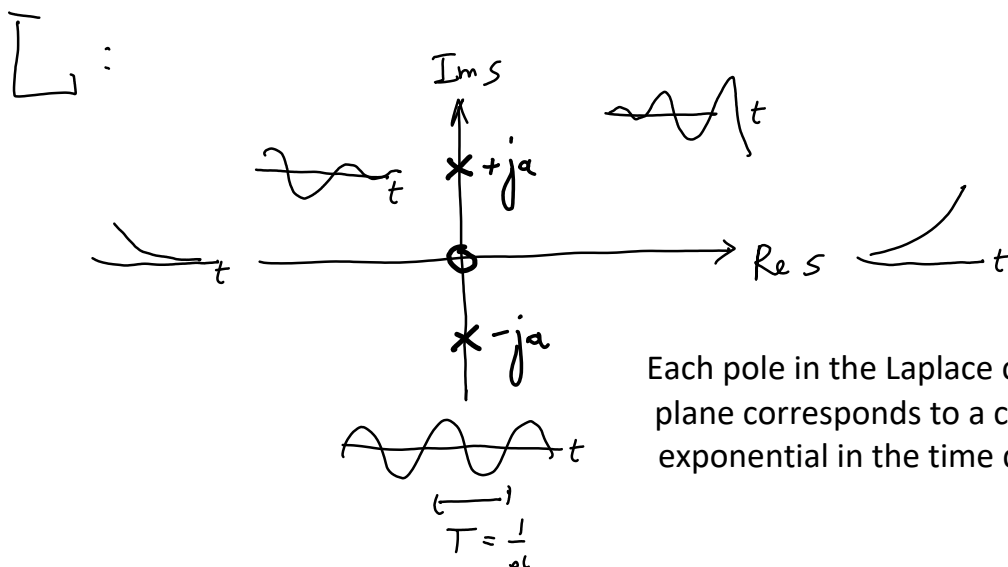
$$\mathcal{L} : s = \sigma + j\omega \quad \begin{array}{l} \text{COMPLEX} \\ \rightarrow \text{exponentials} \end{array}$$

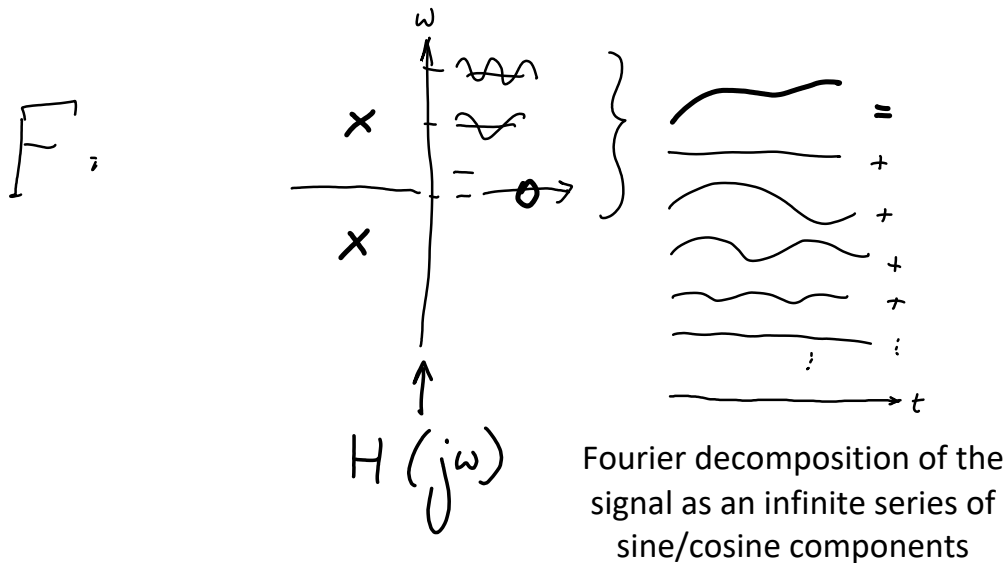
$$\mathcal{F} : s = j\omega \quad \begin{array}{l} \text{PURELY IMAGINARY} \\ \rightarrow \text{sines / cosines} \end{array}$$

$$\mathcal{L} : u(t) = e^{\sigma t} e^{j\omega t} + \text{c.c.}$$

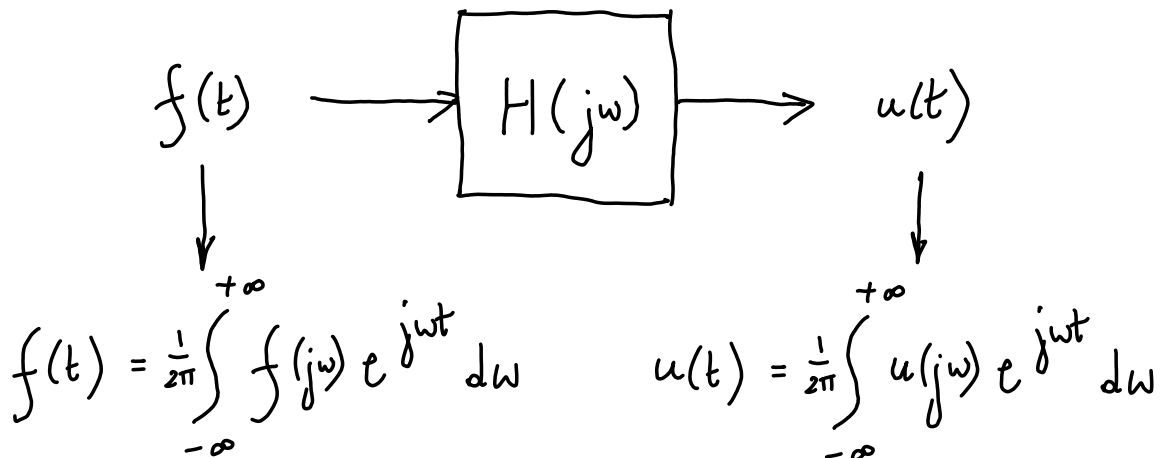
$$\mathcal{F} : u(t) = e^{j\omega t} + \text{c.c.}$$

$\cos(\omega t) + j\sin(\omega t)$       complex conjugate





Frequency response of a linear time-invariant (LTI) system:



Fourier decomposition of the input  $f(t)$  as an infinite series of complex exponentials at radial frequency  $\omega$ , each with complex amplitude  $f(j\omega)$ .

Corresponding Fourier decomposition of the output  $u(t)$ , where each output component  $u(j\omega)$  is linearly related to the input component  $f(j\omega)$  at the *same* radial frequency  $\omega$ :

$$u(j\omega) = H(j\omega) \cdot f(j\omega)$$

The complex gain of this linear relationship is given by the transfer function of the system response at  $s = j\omega$ .

The *magnitude* of this complex gain scales the *amplitude* of the sinusoid from input to output, and the *angle* offsets the *phase* of the sinusoid from input to output:

$$H(j\omega) = A(\omega) \cdot e^{j\varphi(\omega)}$$

Transfer function  
(complex)
Amplitude  
(Gain)

Phase

Single-frequency complex sinusoid:

$$e^{j\omega t} \longrightarrow A(\omega) \cdot e^{j(\omega t + \varphi(\omega))}$$

$$\cos \omega t + j \sin \omega t \longrightarrow A(\omega) \cdot (\cos(\omega t + \varphi(\omega)) + j \sin(\omega t + \varphi(\omega)))$$

Each frequency component:

$$f(j\omega) \longrightarrow A(\omega) \cdot e^{j\varphi(\omega)} \cdot f(j\omega)$$

$$A_f e^{j\varphi_f} \longrightarrow \underbrace{A(\omega) \cdot A_f}_{A_u} e^{j(\underbrace{\varphi(\omega) + \varphi_f}_{\varphi_u})}$$

Amplitudes multiply  
 (but add in their logarithms)

Phases add

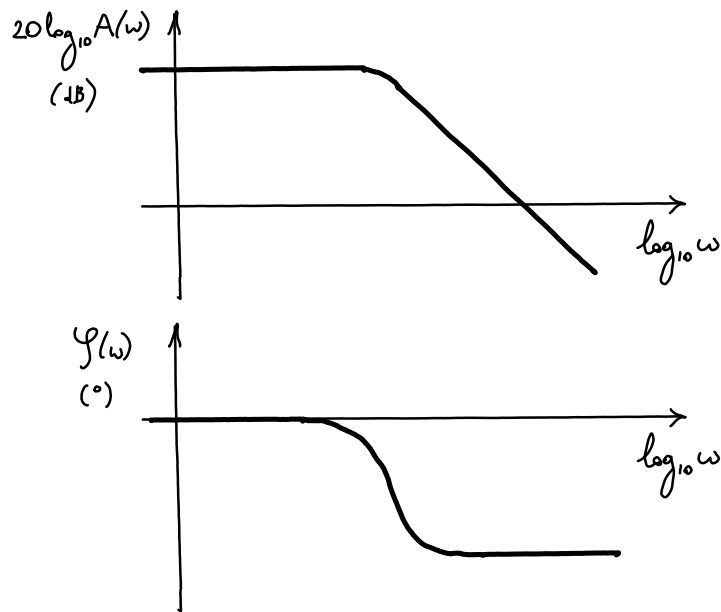
Bode frequency analysis:

**GAIN (dB):**  $20 \log_{10} A_u = 20 \log_{10} A(\omega) + 20 \log_{10} A_f$

**PHASE (°):**  $\underbrace{\varphi_u}_{\text{Output}} = \underbrace{\varphi(\omega)}_{\text{System response}} + \underbrace{\varphi_f}_{\text{Input}}$

--> **Bode plot**

log amplitude & phase as a function of log frequency, e.g.:



Units:

Gain:

$$A(\omega) = |H(j\omega)|$$

[dB] : 20 dB/dec  
decibels 6 dB/oct

Phase:

$$\varphi(\omega) = \angle H(j\omega)$$

[°] degrees  
[rad] radians

Radial/cyclic frequency:

$$\omega = 2\pi f$$

$\left[\frac{\text{rad}}{\text{s}}\right]$       [Hz]